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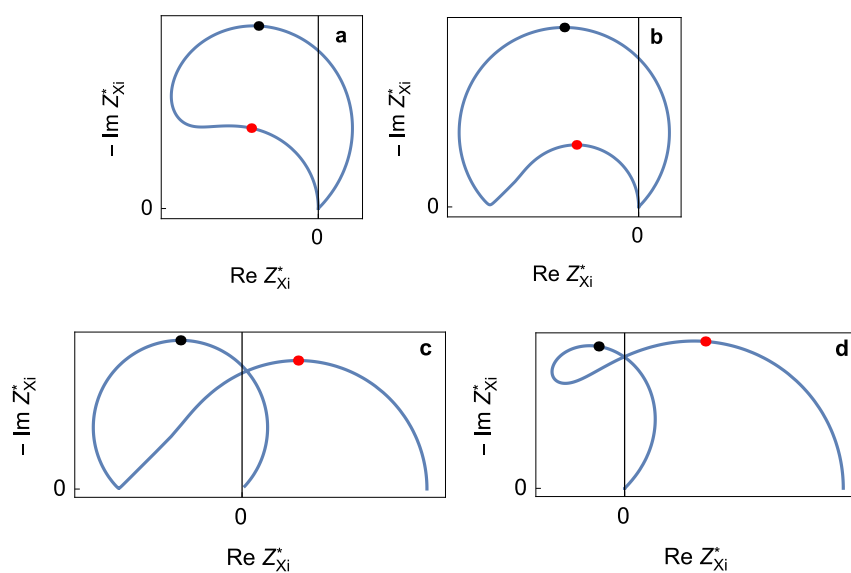


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Handbook of Electrochemical Impedance Spectroscopy



DISTRIBUTED and MIXED IMPEDANCES

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J.-P. Diard, B. Le Gorrec, C. Montella

July 16, 2020

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Chapter 1

Introduction

1.1 Lumped *vs.* distributed systems

1.1.1 Lumped systems

The transfer functions of systems modeled by ordinary differential equations, often called lumped-parameter systems, are rational functions (i.e. a ratio of two polynomials in s , the Laplace variable) [1, 2].

1.1.2 Distributed systems

The transfer functions of distributed parameter systems are irrational functions. The analysis of rational and irrational transfer functions differ in a number of important aspects. The most obvious differences between rational and irrational transfer functions are the poles and zeros. Irrational transfer functions often have infinitely many poles and zeros [2].

1.1.3 Mixed lumped-distributed systems

The transfer functions of mixed lumped-distributed systems contain rational and irrational functions in s .

1.2 Examples in electrochemistry

1.2.1 Lumped systems

Faradaic impedance Z_f and impedance Z of electrochemical adsorption reaction (EAR) are lumped systems [3, 4]. Eq. (1.1) is a rational fraction in s ⁽¹⁾.

$$Z(s) = \frac{1 + R_{ct}C_{ads}s}{s((C_{dl} + C_{ads}) + sR_{ct}C_{dl}C_{ads})} \quad (1.1)$$

$$Z(s) = K \frac{1 + \alpha s}{s(1 + \beta s)}, K = \frac{1}{C_{dl} + C_{ads}}, \alpha = R_{ct}C_{ads}, \beta = \frac{R_{ct}C_{dl}C_{ads}}{C_{dl} + C_{ads}} \quad (1.2)$$

¹Replacing a capacitor, for example C_{dl} , by a CPE [5] transforms a lumped impedance in a distributed impedance. This case is not subsequently envisaged.

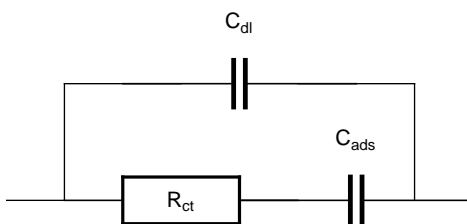


Figure 1.1: Equivalent circuit for electrochemical adsorption reaction (EAR).

1.2.2 Distributed systems

The Faradaic impedance of a corroding electrode with mass transfer limitation (Fig 1.2) is a rational fraction in \sqrt{s} , i.e. an irrational fraction in s .

$$Z_f(s) = \frac{R\sigma}{\sigma + R\sqrt{s}} \quad (1.3)$$

$$Z_f(s) = \frac{K}{1 + \alpha\sqrt{s}}, \quad K = R, \quad \alpha = \frac{R}{\sigma} \quad (1.4)$$

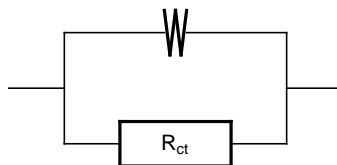


Figure 1.2: Equivalent circuit for a corroding electrode with mass transfer limitation. W: Warburg element for semi-infinite linear diffusion [6].

1.2.3 Mixed lumped-distributed systems

The impedance of the Randles equivalent circuit [4, 6, 7] (Fig. 1.3) is a mixed lumped and distributed system:

$$Z(s) = \frac{R_{ct} + R_d \frac{\text{th} \sqrt{\tau_d s}}{\sqrt{\tau_d s}}}{1 + R_{ct} C_{dl} s + C_{dl} s R_d \frac{\text{th} \sqrt{\tau_d s}}{\sqrt{\tau_d s}}} \quad (1.5)$$

$$Z(s) = K \frac{1 + \alpha \frac{\text{th} \sqrt{\tau_d s}}{\sqrt{\tau_d s}}}{1 + \beta s + \gamma s \frac{\text{th} \sqrt{\tau_d s}}{\sqrt{\tau_d s}}}, \quad K = R_{ct}, \quad \alpha = \frac{R_d}{R_{ct}}, \quad \beta = R_{ct} C_{dl} \quad \gamma = C_{dl} R_d \quad (1.6)$$

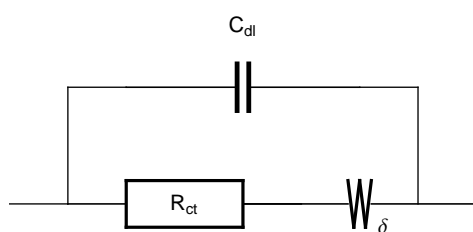


Figure 1.3: Randles equivalent circuit for a redox reactions studied on a rotating disk electrode. W_{δ} : bounded diffusion impedance [6].

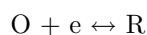
Chapter 2

Impedance containing $\frac{\text{th } \sqrt{S}}{\sqrt{S}}$

2.1 $\frac{\text{th } \sqrt{S}}{\sqrt{S}}$

2.1.1 Electrochemical reaction

Redox reaction [4, 8, 9]:



studied on a rotating disk electrode with mass transfer limitation.

2.1.2 Electrochemical impedance

$$Z_{W_\delta}(s) = R_d \frac{\text{th } \sqrt{\tau s}}{\sqrt{\tau s}} \quad (1) \quad (2.1)$$

2.1.3 Reduced impedance

$$Z_{W_\delta}(s) = R_d \frac{\text{th } \sqrt{\tau s}}{\sqrt{\tau s}} \Rightarrow Z(S) = \frac{Z_{W_\delta}(s)}{R_d} = \frac{\text{th } \sqrt{S}}{\sqrt{S}}, \quad S = \tau s = \Sigma + i u \quad (2.2)$$

2.1.4 Pole-zero map

Infinite product expansion [12–16]:

$$\frac{\text{th } \sqrt{S}}{\sqrt{S}} = \frac{1}{1 + \frac{4S}{\pi^2}} \prod_{k=1}^{\infty} \frac{1 + \frac{S}{(k\pi)^2}}{1 + \frac{4S}{((2k+1)\pi)^2}} = P_\infty \quad (2.3)$$

Thanks to Eq. (2.3)

¹This expression could be replaced by a more accurate one [10, 11]. This case is not subsequently envisaged.

- infinity of interlaced real poles and zeros (Fig. 2.1).

$$s_p = -\frac{1}{4}((2k+1)\pi)^2, \quad k = 1 \cdots \infty \quad (2.4)$$

$$s_z = -(k\pi)^2, \quad k = 1 \cdots \infty \quad (2.5)$$

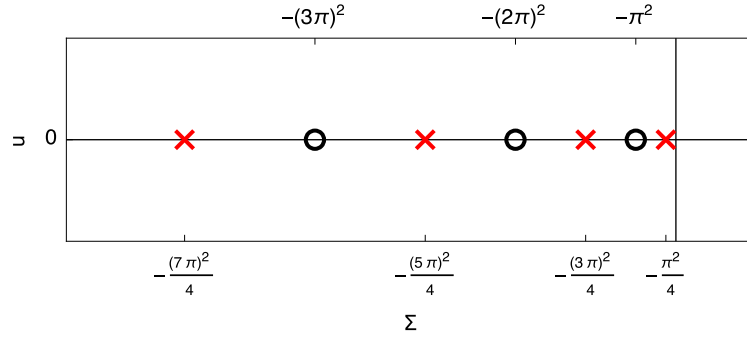


Figure 2.1: Pole-zero map of $\frac{\text{th}\sqrt{S}}{\sqrt{S}}$.

2.1.5 Graph of the reduced impedance

3D plot of the modulus (Fig. 2.2).

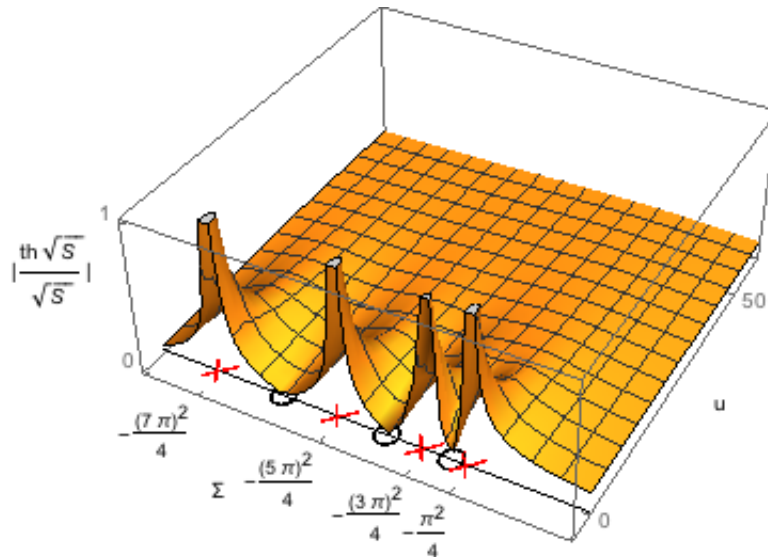


Figure 2.2: 3D plot of the modulus of $\frac{\text{th}\sqrt{S}}{\sqrt{S}}$.

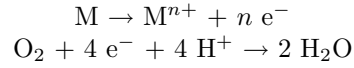
$$2.2. \frac{\frac{TH\sqrt{S}}{\sqrt{S}}}{1 + \alpha \frac{TH\sqrt{S}}{\sqrt{S}}}$$

$$2.2 \frac{\frac{\text{th}\sqrt{S}}{\sqrt{S}}}{1 + \alpha \frac{\text{th}\sqrt{S}}{\sqrt{S}}}$$

11

2.2.1 Electrochemical reaction

Corroding electrode with mass transfert limitation:



2.2.2 Reduced Faradaic impedance

$$Z(S) = \frac{\frac{\text{th}\sqrt{S}}{\sqrt{S}}}{1 + \alpha \frac{\text{th}\sqrt{S}}{\sqrt{S}}}, \quad u_{c1} = 2.54, u_{c2} = \alpha^2 \quad (2.6)$$

2.2.3 Nyquist diagrams

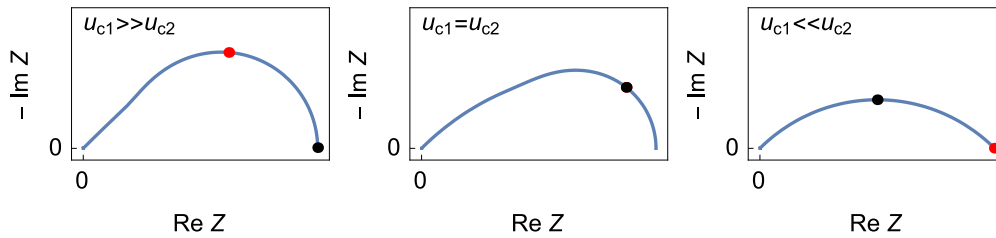


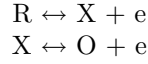
Figure 2.3: Nyquist diagrams calculated from Eq. (2.6). Red dots : $u_{c1} = 2.54$, black dots : $u_{c2} = \alpha^2$.

- $u_{c1} \gg u_{c2}, 2.54 \gg \alpha^2 \Rightarrow$ quarter of a lemniscate,
- $u_{c1} \ll u_{c2}, 2.54 \ll \alpha^2 \Rightarrow$ quarter of a circle (see Annex A.2).

$$2.3 \quad \frac{\left(1 + \alpha \frac{\text{th}\sqrt{S}}{\sqrt{S}}\right) \frac{\text{th}\sqrt{S}}{\sqrt{S}}}{1 + \beta \frac{\text{th}\sqrt{S}}{\sqrt{S}}}$$

2.3.1 Electrochemical reaction

EE reaction [17]:



studied on a rotating disk electrode with $D_{\text{R}} = D_{\text{X}} = D_{\text{O}}$.

2.3.2 Reduced concentration impedance

Concentration impedances of soluble species:

$$Z_{X_i}(S) = \frac{\left(1 + \alpha \frac{\text{th}\sqrt{S}}{\sqrt{S}}\right) \frac{\text{th}\sqrt{S}}{\sqrt{S}}}{1 + \beta \frac{\text{th}\sqrt{S}}{\sqrt{S}}}, \quad \alpha, \beta \geq 0 \quad (2.7)$$

2.3.3 Nyquist diagrams

Figs. 2.4, 2.5 and 2.6.

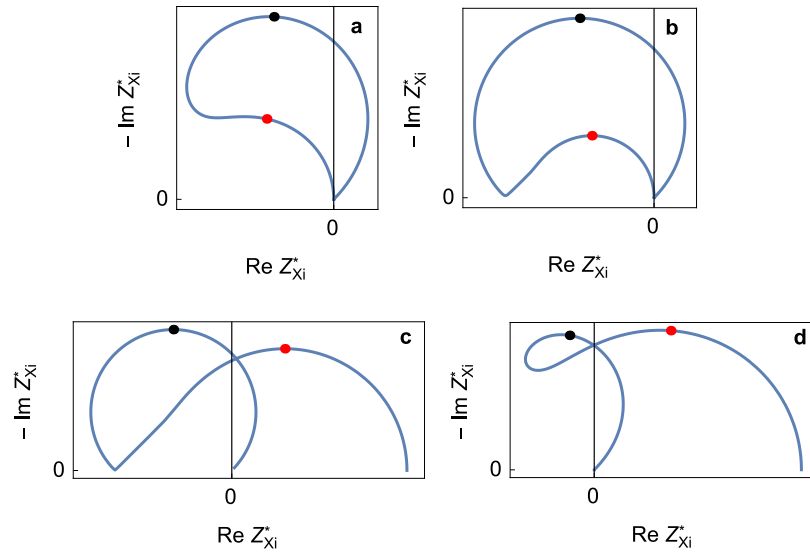


Figure 2.4: Some amazing Nyquist diagrams calculated from Eq. (2.7), $\alpha = -1$, $\beta = -10$ (left), $\beta = -10^4$ (right). Red dots : $u_{c1} = 2.54$, black dots : $u_{c2} = \beta^2$. a: $\alpha = -1$, $\beta = -10$, b: $\alpha = -1$, $\beta = -10^{-4}$, c: $\alpha = -2.5$, $\beta = -10^{-5}$, d: $\alpha = -2.5$, $\beta = -10$ (Hokusai's great wave).

$$2.3. \frac{\left(1 + \alpha \frac{TH\sqrt{S}}{\sqrt{S}}\right) \frac{TH\sqrt{S}}{\sqrt{S}}}{1 + \beta \frac{TH\sqrt{S}}{\sqrt{S}}}$$

$$\frac{(1 + \alpha M(S)) M(S)}{1 + \beta M(S)}, M(S) = \frac{th \sqrt{S}}{\sqrt{S}}$$

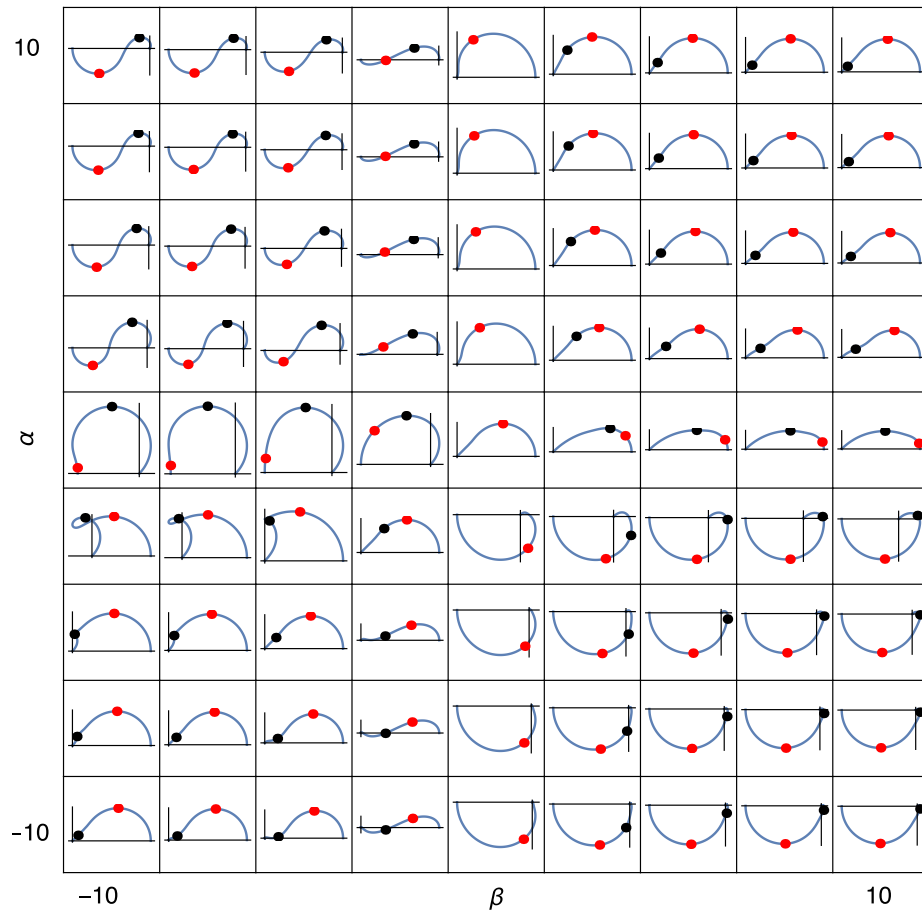


Figure 2.5: Array of impedance diagrams calculated from Eq. (2.7). Red dots : $u_{c1} = 2.54$, black dots : $u_{c2} = \beta^2$.

$$\frac{(1 + \alpha M(S)) M(S)}{1 + \beta M(S)}, M(S) = \frac{\text{th}\sqrt{S}}{\sqrt{S}}$$

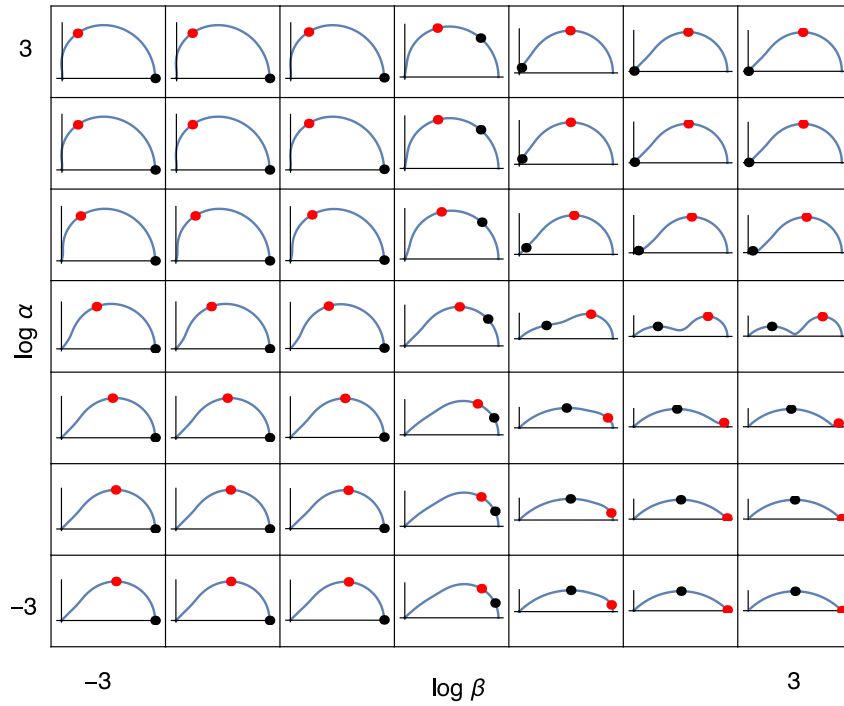


Figure 2.6: Array of impedance diagrams calculated from Eq. (2.7). Red dots : $u_{c1} = 2.54$.

Chapter 3

Impedance containing lumped and distributed elements

$$3.1 \quad \frac{\operatorname{th} \sqrt{S}}{\sqrt{S}} \\ 1 + \alpha S$$

$$Z(S) = \frac{\operatorname{th} \sqrt{S}}{1 + \alpha S} \quad (3.1)$$

Characteristic frequencies:

- $u_{c1} = 2.54$
- $u_{c2} = 1/\alpha$

3.1.1 Pole-zero map

- Same zeros as $\frac{\operatorname{th} \sqrt{S}}{\sqrt{S}}$
- Same poles as $\frac{\operatorname{th} \sqrt{S}}{\sqrt{S}}$ plus one real pole $(-\frac{1}{\alpha})$ (Figs. 3.4-3.1).

3.1.2 Nyquist diagrams

Figs. 3.4-3.1.

- $u_{c1} \ll u_{c2} \Rightarrow Z(S) \approx \frac{\operatorname{th} \sqrt{S}}{\sqrt{S}}$
- $u_{c1} \gg u_{c2} \Rightarrow Z(S) \approx \frac{1}{1 + \alpha S}$

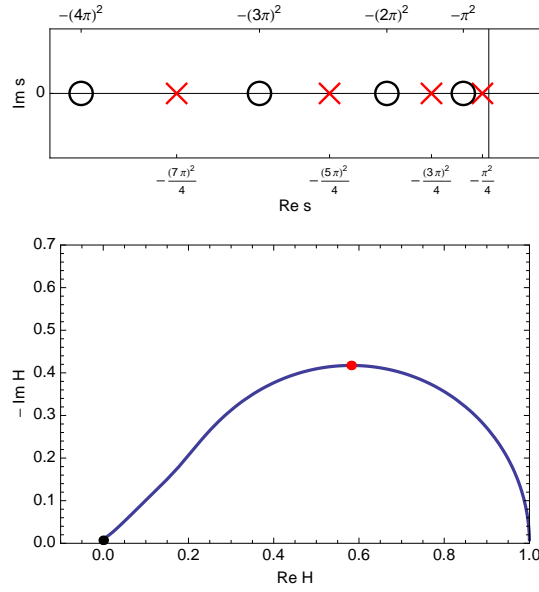


Figure 3.1: Pole-zero map and Nyquist diagram of $\frac{1}{1 + \alpha S} \frac{\text{th} \sqrt{S}}{\sqrt{S}}$. $\alpha = 10^{-4}$.
Red dot: $u_{c1} = 2.54$, black dot: $u_{c2} = 1/\alpha$.

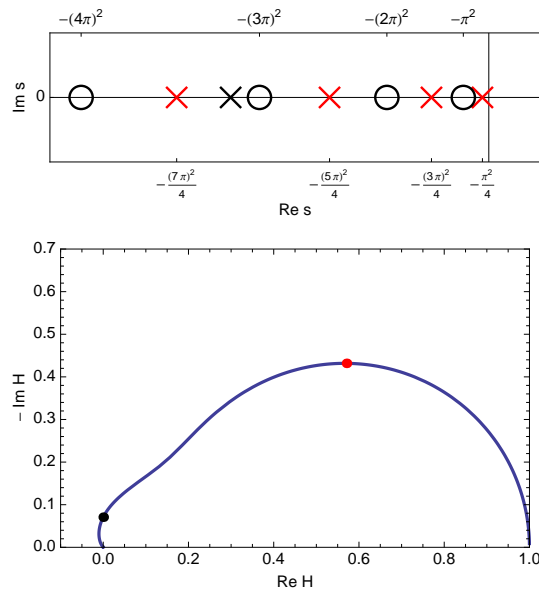


Figure 3.2: Pole-zero map and Nyquist diagram of $\frac{1}{1 + \alpha S} \frac{\text{th} \sqrt{S}}{\sqrt{S}}$. $\alpha = 10^{-2}$.
Red dot: $u_{c1} = 2.54$, black dot: $u_{c2} = 1/\alpha$.

$$3.1. \frac{\frac{TH\sqrt{S}}{\sqrt{S}}}{1 + \alpha S}$$

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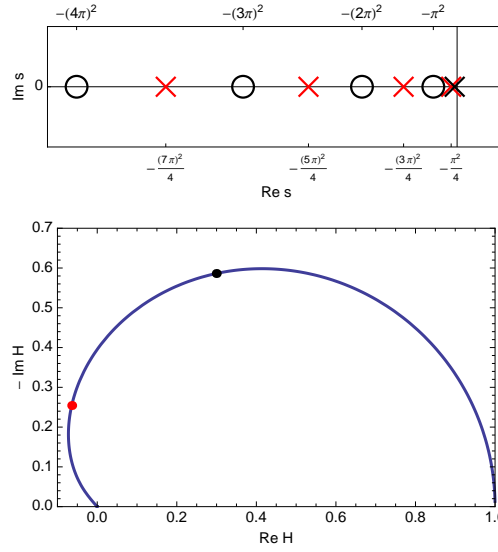


Figure 3.3: Pole-zero map and Nyquist diagram of $\frac{1}{1 + \alpha S} \frac{\text{th}\sqrt{S}}{\sqrt{S}}$. $\alpha = 1$ ($-\frac{1}{\alpha} > -\frac{\pi^2}{4}$). Red dot: $u_{c1} = 2.54$, black dot: $u_{c2} = 1/\alpha$.

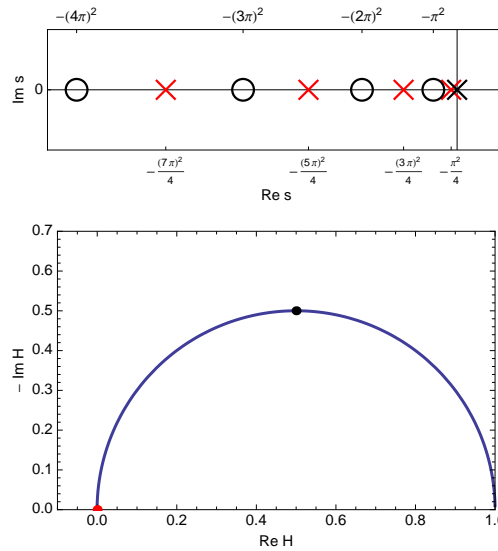


Figure 3.4: Pole-zero map and Nyquist diagram of $\frac{1}{1 + \alpha s} \frac{\text{th}\sqrt{S}}{\sqrt{S}}$. $\alpha = 10^3$ ($-\frac{1}{\alpha} > -\frac{\pi^2}{4}$). Red dot: $u_{c1} = 2.54$, black dot: $u_{c2} = 1/\alpha$.

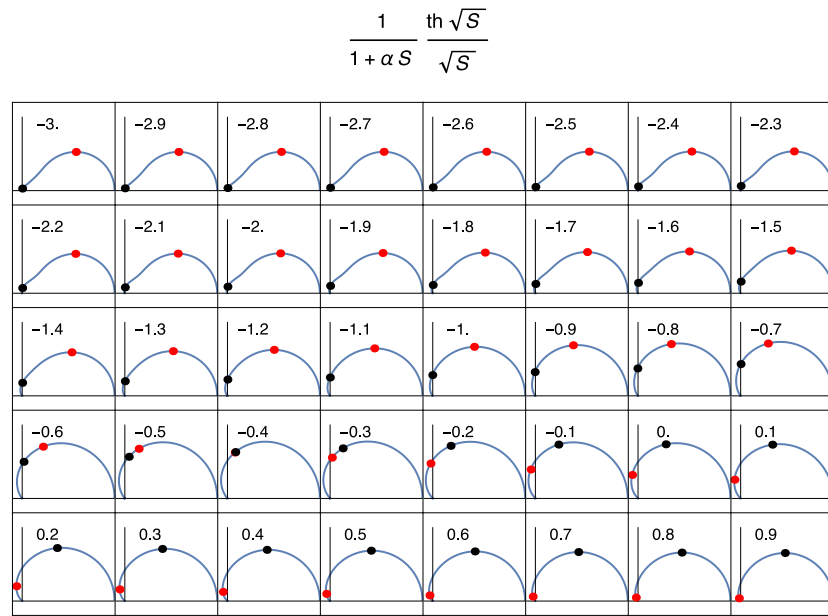


Figure 3.5: Change of Nyquist diagram of $\frac{1}{1 + \alpha S} \frac{\text{th} \sqrt{S}}{\sqrt{S}}$ with increasing value of α from 10^{-3} to $10^{0.9}$. Decimal logarithm of α reported on the Nyquist diagrams. Red dots: $u_{c1} = 2.54$, black dots: $u_{c2} = 1/\alpha$.

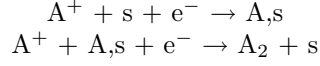
$$3.2. \frac{1 + \alpha \frac{\text{th} \sqrt{S}}{\sqrt{S}}}{1 + \beta S}$$

19

$$3.2 \quad \frac{1 + \alpha \frac{\text{th} \sqrt{S}}{\sqrt{S}}}{1 + \beta S}$$

3.2.1 Electrochemical reaction: Volmer-Heyrovský (V-H)

Electrochemical reaction: Volmer-Heyrovský (V-H) [4, 8]



3.2.2 Reduced concentration impedance of adsorbed species

$$Z_\theta^*(S) = \frac{1 + \alpha \frac{\text{th} \sqrt{S}}{\sqrt{S}}}{1 + \beta S} = \frac{1}{1 + \beta S} + \frac{\alpha \frac{\text{th} \sqrt{S}}{\sqrt{S}}}{1 + \beta S} \quad (3.2)$$

$$\alpha \rightarrow \infty \Rightarrow Z_\theta^*(S) \approx \alpha \frac{\text{th} \sqrt{S}}{\sqrt{S}} \Rightarrow u_{c1} = 2.54 \quad (3.3)$$

$$\alpha \rightarrow 0 \Rightarrow Z_\theta^*(S) \approx \frac{1}{1 + \beta i u} \Rightarrow u_{c2} = \frac{1}{\beta} \quad (3.4)$$

3.2.3 Nyquist diagrams

Figs. 3.6 and 3.7.

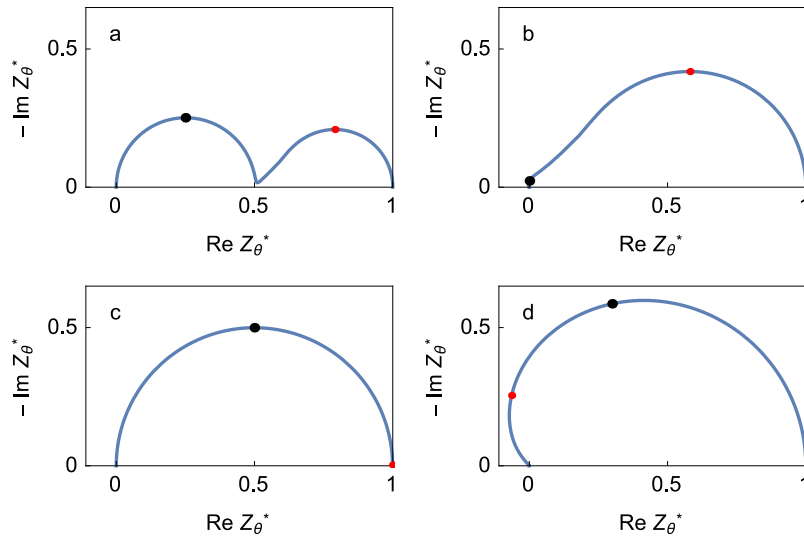


Figure 3.6: Impedance diagrams calculated from Eq. (3.2). a : $\alpha = 1$, $\beta = 10^{-5}$, b : $\alpha = 10^3$, $\beta = 10^{-3}$, c : $\alpha = 10^{-3}$, $\beta = 10^{-3}$, d : $\alpha = 10^3$, $\beta = 1$. Characteristic dimensionless frequencies: red dots : $u_{c1} = 2.54$, black dots : $u_{c2} = 1/\beta$.

$$\frac{1 + \alpha M(s)}{1 + \beta s}, M(s) = \frac{\text{th} \sqrt{s}}{\sqrt{s}}$$

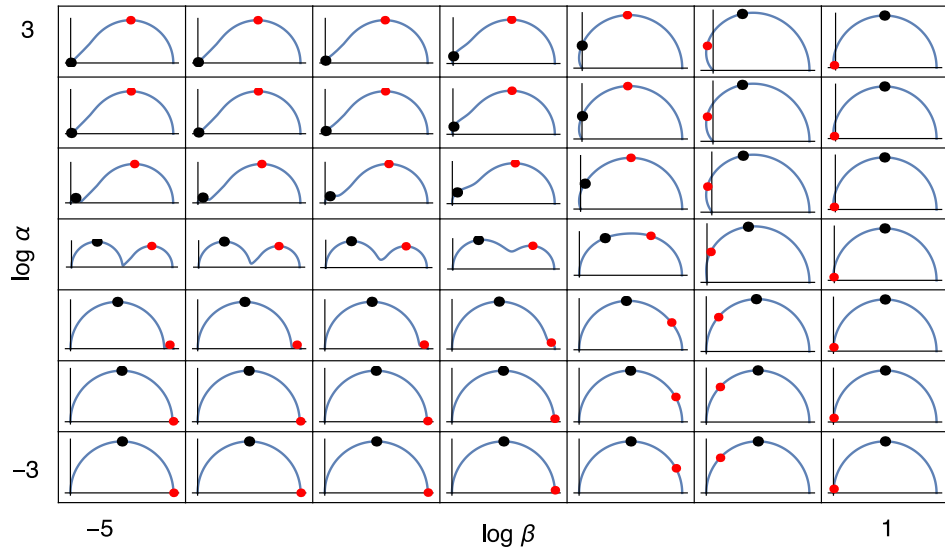


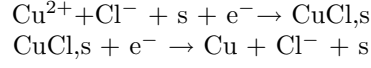
Figure 3.7: Array of impedance diagrams calculated from Eq. (3.2). Impedance diagrams are made of one or two arcs. Characteristic dimensionless frequencies: red dots : $u_{c1} = 2.54$, black dots : $u_{c2} = 1/\beta$.

$$3.3. \frac{1}{1 + \alpha S + \beta S \frac{\text{th} \sqrt{S}}{\sqrt{S}}}$$

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$$3.3 \frac{1}{1 + \alpha S + \beta S \frac{\text{th} \sqrt{S}}{\sqrt{S}}}$$

3.3.1 Electrochemical reaction: catalytic copper deposition



Hypotheses: no mass transfer limitation for Cu^{2+} , ($Cu^{2+}(0, t) \approx Cu^{2+}$), kinetic irreversibility of the two steps [18, 19].

3.3.2 Reduced concentration impedance of adsorbed species

$$Z(S) = \frac{1}{1 + \alpha S + \beta S \frac{\text{th} \sqrt{S}}{\sqrt{S}}} \quad (3.5)$$

- Poles and zeros of $Z^*(S)$ are real and interlaced.
- Zeros of $Z(S)$ are the poles of $\frac{\text{th} \sqrt{S}}{\sqrt{S}}$: $-\frac{1}{4}((2k+1)\pi)^2$, $k = 1 \cdots \infty$.
- Characteristic dimensionless frequencies: $u_{c1} = 2.54$, $u_{c2} = 1/\alpha$, $u_{c3} = 1/\beta$

3.3.3 Nyquist diagrams

Figs. 3.8 and 3.9.

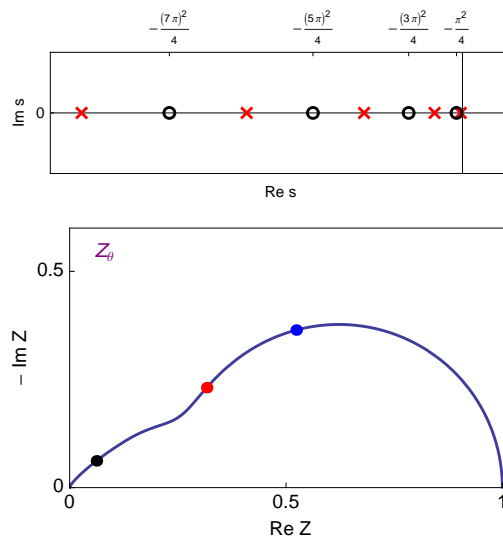


Figure 3.8: Pole-zero map and impedance diagrams calculated from Eq. (3.5). $\alpha = 10^{-2}$, $\beta = 1$, red dot : $u_{c1} = 2.54$, black dot: $u_{c2} = 1/\alpha$, blue dot: $u_{c3} = 1/\beta$.

3.3. $\frac{1}{1 + \alpha S + \beta S \frac{\text{th} \sqrt{S}}{\sqrt{S}}}$

$$\frac{1}{1 + \alpha S + \beta S M(S)}, M(S) = \frac{\text{th} \sqrt{S}}{\sqrt{S}}$$

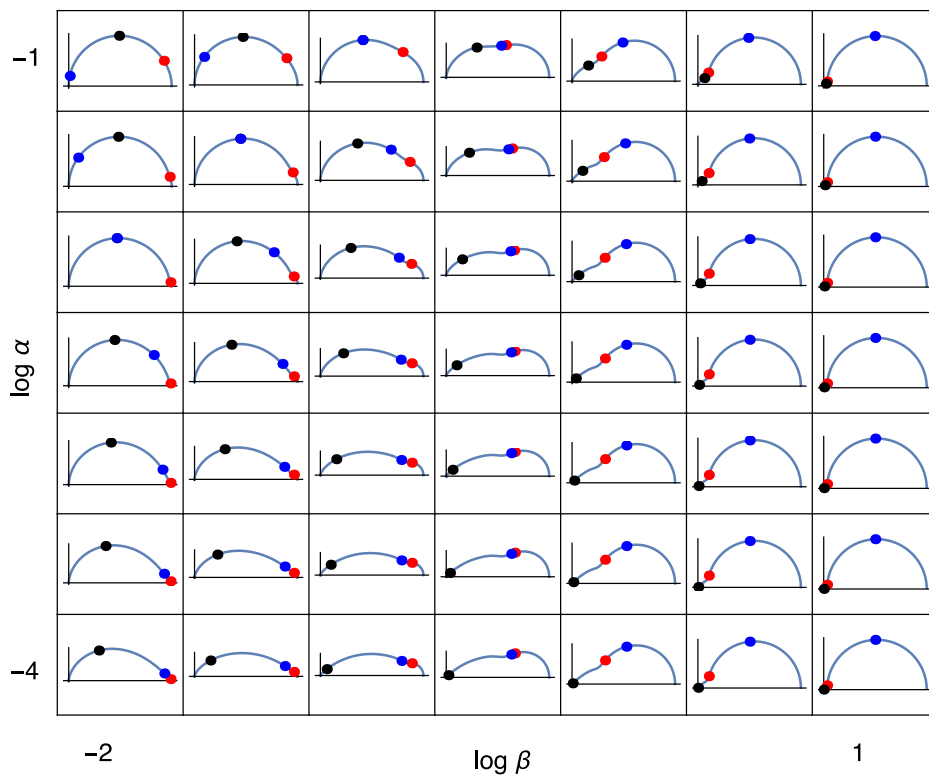
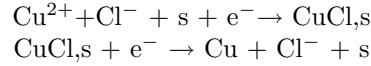


Figure 3.9: Graphics array representation of the impedance diagram, calculated from Eq. (3.5) and plotted using the Nyquist representation (orthonormal scales) for catalytic copper deposition. Characteristic dimensionless frequencies: red dots : $u_{c1} = 2.54$, black dots: $u_{c2} = 1/\alpha$, blue dots: $u_{c3} = 1/\beta$.

$$3.4 \quad \frac{S \frac{\text{th} \sqrt{S}}{\sqrt{S}}}{1 + \alpha S + \beta S \frac{\text{th} \sqrt{S}}{\sqrt{S}}}$$

3.4.1 Electrochemical reaction: catalytic copper deposition



Hypotheses: no mass transfer limitation for Cu^{2+} , ($Cu^{2+}(0, t) \approx Cu^{2+}$), kinetic irreversibility of the two steps [19].

3.4.2 Reduced concentration impedance of soluble species Cl^-

$$Z(S) = \frac{S \frac{\text{th} \sqrt{S}}{\sqrt{S}}}{1 + \alpha S + \beta S \frac{\text{th} \sqrt{S}}{\sqrt{S}}} \quad (3.6)$$

3.4.3 Nyquist diagrams

- Poles and zeros of $Z(S)$ are real.
- Zeros of $Z(S)$ are the zeros of $\frac{\text{th} \sqrt{S}}{\sqrt{S}}$ ($S_Z = -(k \pi)^2$, $k = 1 \cdots \infty$) plus one zero at the origine of the complex plane (derivator).
- Characteristic dimensionless frequencies: $u_{c1} = 2.54$, $u_{c2} = 1/\alpha$, $u_{c3} = 1/\beta$, $u_{c4} = (\beta/\alpha)^2$.

Fig. 3.10.

$$3.4. \frac{S \frac{TH\sqrt{S}}{\sqrt{S}}}{1 + \alpha S + \beta S \frac{TH\sqrt{S}}{\sqrt{S}}}$$

25

$$\frac{SM(S)}{1 + \alpha S + \beta SM(S)}, M(S) = \frac{\text{th } \sqrt{S}}{\sqrt{S}}$$

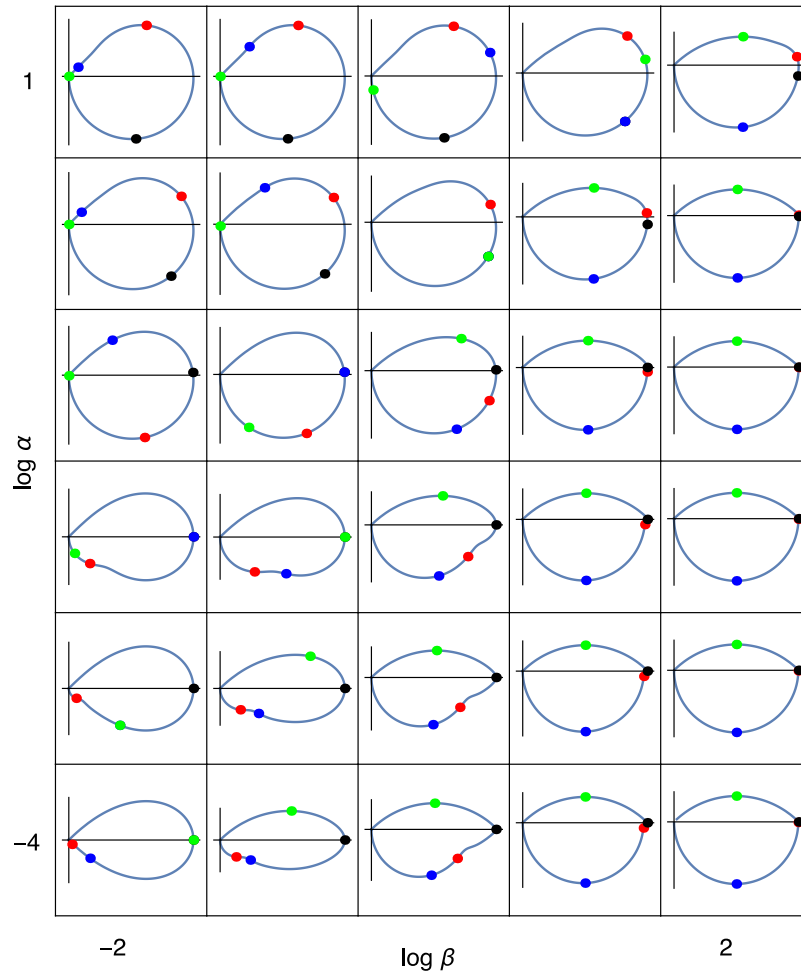
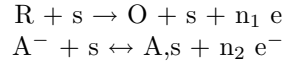


Figure 3.10: Graphics array representation of the reduced impedance diagram calculated from Eq. (3.6) and plotted using the Nyquist complex plane representation (orthonormal scales) for catalytic copper deposition. Characteristic dimensionless frequencies: red dots : $u_{c1} = 2.54$, black dots: $u_{c2} = 1/\alpha$, blue dots: $u_{c3} = 1/\beta$, green dots: $u_{c4} = (\beta/\alpha)^2$.

$$3.5 \quad \frac{1 + \alpha \frac{\text{th} \sqrt{S}}{\sqrt{S}}}{1 + \beta S + \gamma S \frac{\text{th} \sqrt{S}}{\sqrt{S}}}$$

3.5.1 Electrochemical reaction: E-EAR reaction [20]



3.5.2 Reduced concentration impedance of adsorbed species

$$Z(S) = \frac{1 + \alpha \frac{\text{th} \sqrt{S}}{\sqrt{S}}}{1 + \beta S + \gamma S \frac{\text{th} \sqrt{S}}{\sqrt{S}}}, \quad \alpha, \beta, \gamma \geq 0 \quad (3.7)$$

3.5.3 Nyquist diagrams

$$\frac{1 + \alpha M(S)}{1 + \beta S + \gamma S M(S)}, \quad M(S) = \frac{\text{th} \sqrt{S}}{\sqrt{S}}, \quad \gamma = 0.01$$

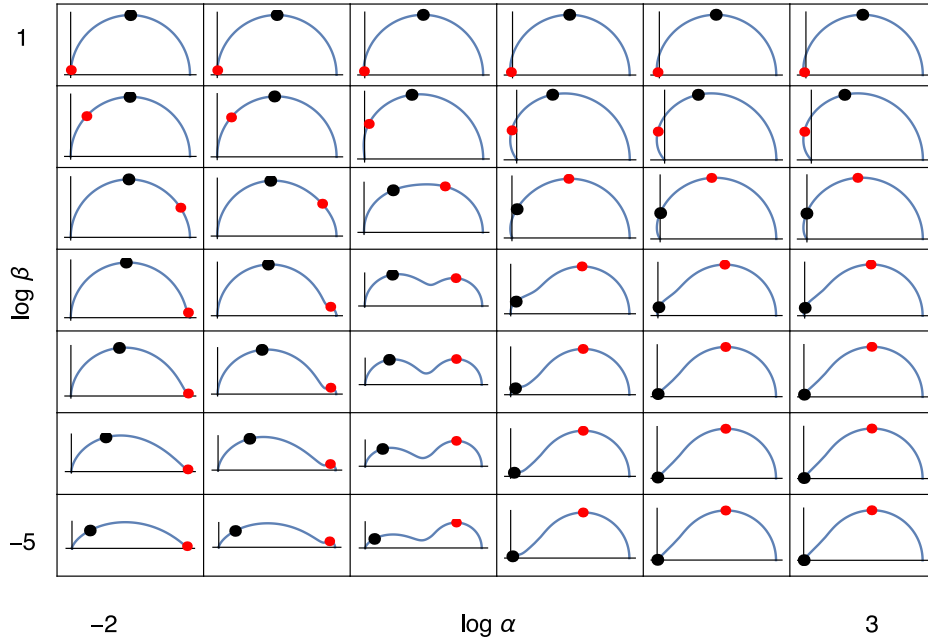


Figure 3.11: Graphics array representation of the Nyquist diagram for the impedance calculated from Eq. (3.7) and plotted using the Nyquist representation (orthonormal scales). Characteristic dimensionless angular frequencies: red dots: $u_{c1} = 2.54$, black dots: $u_{c2} = 1/\beta_1$. $\gamma = 10^{-2}$.

$$3.5. \frac{1 + \alpha \frac{TH\sqrt{S}}{\sqrt{S}}}{1 + \beta S + \gamma S \frac{TH\sqrt{S}}{\sqrt{S}}}$$

$$\frac{1 + \alpha M(S)}{1 + \beta S + \gamma S M(S)}, M(S) = \frac{th \sqrt{S}}{\sqrt{S}}, \gamma = -0.1$$

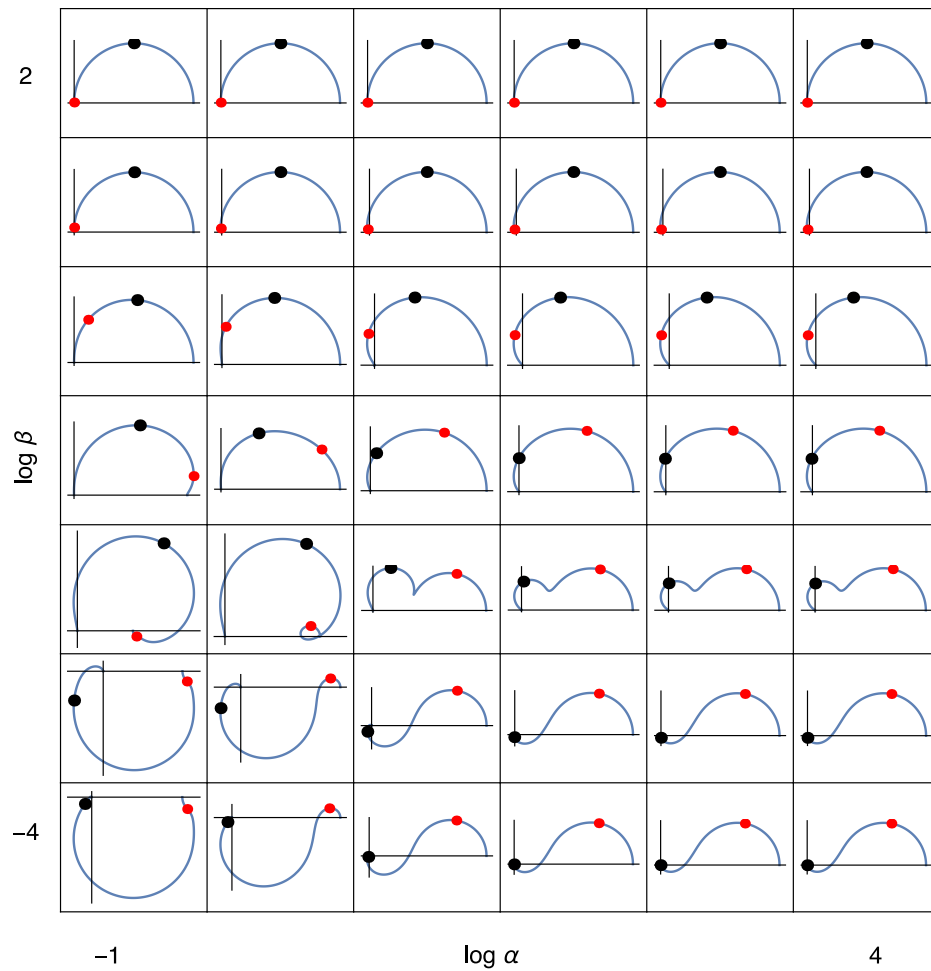


Figure 3.12: Graphics array representation of the Nyquist diagram for the impedance calculated from Eq. (3.7) and plotted using the Nyquist representation (orthonormal scales). Characteristic dimensionless angular frequencies: red dots: $u_{c1} = 2.54$, black dots: $u_{c2} = 1/\beta_1$. $\gamma = -10^{-1}$.

$$\frac{1 + \alpha M(s)}{1 + \beta s + \gamma s M(s)}, M(s) = \frac{\text{th} \sqrt{s}}{\sqrt{s}}, \gamma = -0.1$$

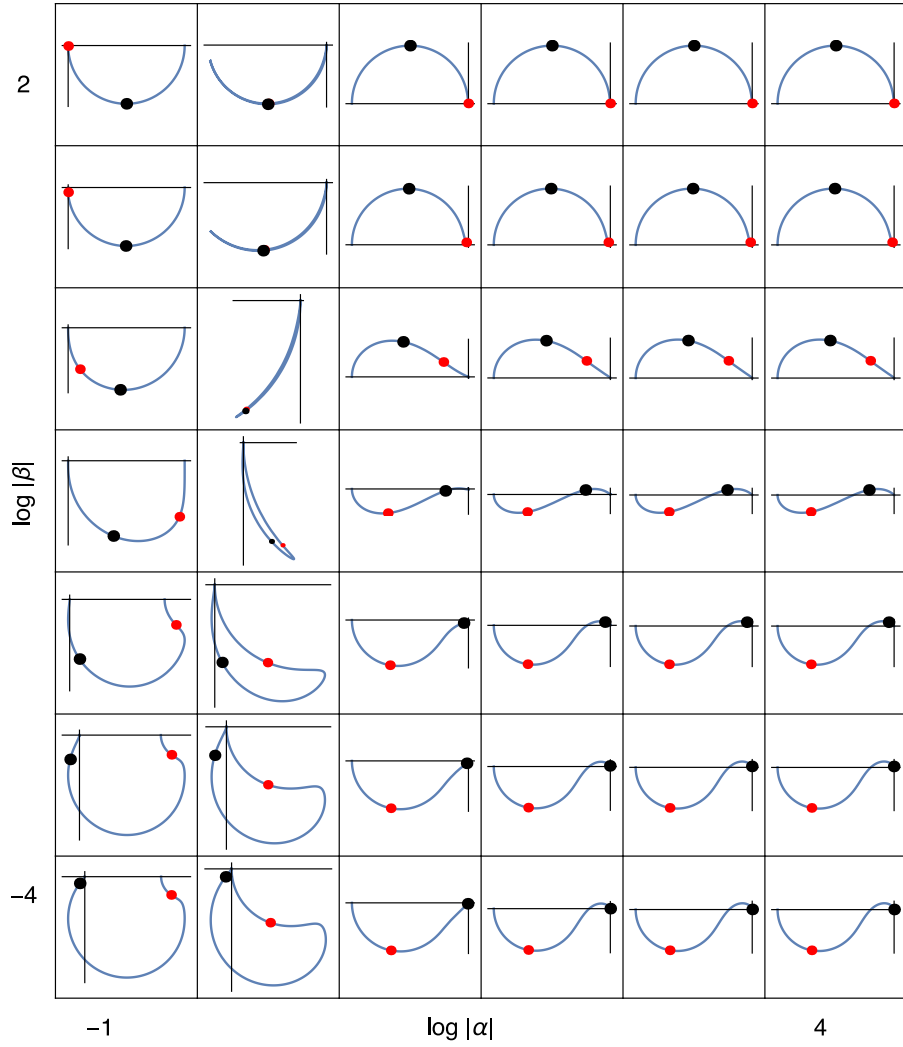
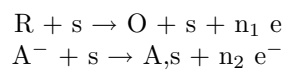


Figure 3.13: Graphics array representation of the Nyquist diagram for the impedance calculated from Eq. 3.7 and plotted using the Nyquist representation (orthonormal scales). $\alpha, \beta, \gamma < 0$. Characteristic dimensionless angular frequencies: red dots: $u_{c1} = 2.54$, black dots: $u_{c2} = 1/|\beta_1|$. $\alpha, \beta, \gamma < 0, \gamma = -10^{-1}$.

$$3.6. \frac{(1 + \alpha S) \frac{\text{th} \sqrt{S}}{\sqrt{S}}}{1 + \beta S + \gamma S \frac{\text{th} \sqrt{S}}{\sqrt{S}}} \quad 29$$

$$\mathbf{3.6} \quad \frac{(1 + \alpha S) \frac{\text{th} \sqrt{S}}{\sqrt{S}}}{1 + \beta S + \gamma S \frac{\text{th} \sqrt{S}}{\sqrt{S}}}$$

3.6.1 Electrochemical reaction: E-EAR reaction [20]



3.6.2 Reduced concentration impedance of soluble species R

$$Z(S) = \frac{(1 + \alpha S) \frac{\text{th} \sqrt{S}}{\sqrt{S}}}{1 + \beta S + \gamma S \frac{\text{th} \sqrt{S}}{\sqrt{S}}}, \quad \alpha, \beta, \gamma \geq 0 \quad (3.8)$$

3.6.3 Nyquist diagrams

Figs. 3.14 and 3.15.

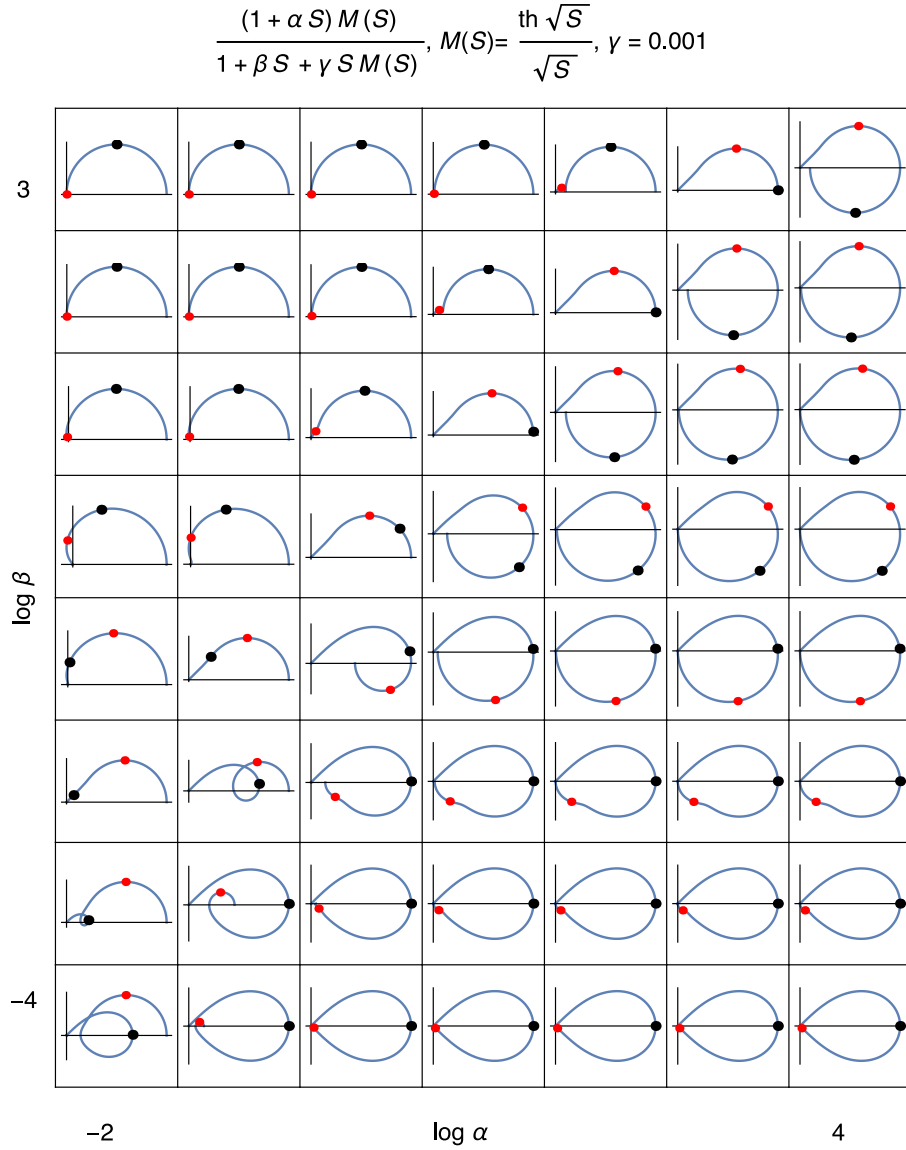


Figure 3.14: Graphics array representation of the Nyquist diagram for the impedance calculated from Eq. (3.8) and plotted using the Nyquist representation (orthonormal scales). Characteristic dimensionless angular frequencies: red dots: $u_{c1} = 2.54$, black dots: $u_{c2} = 1/\beta$. $\gamma = 10^{-3}$.

$$3.6. \frac{(1 + \alpha S) \frac{TH\sqrt{S}}{\sqrt{S}}}{1 + \beta S + \gamma S \frac{TH\sqrt{S}}{\sqrt{S}}}$$

$$\frac{(1 + \alpha S) M(S)}{1 + \beta S + \gamma S M(S)}, M(S) = \frac{\text{th } \sqrt{S}}{\sqrt{S}}, \gamma = 0.001$$

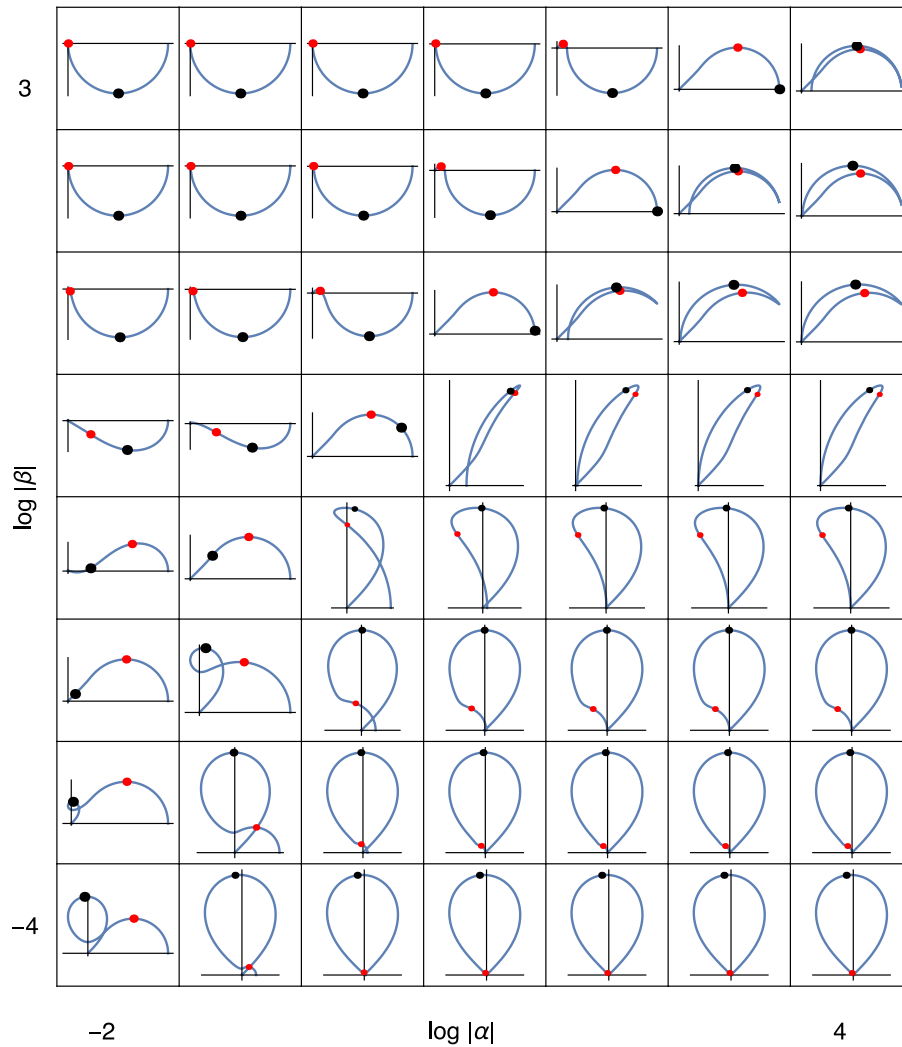


Figure 3.15: Graphics array representation of the Nyquist diagram for the impedance calculated from Eq. (3.8) and plotted using the Nyquist representation (orthonormal scales). $\alpha, \beta, \gamma < 0$. Characteristic dimensionless angular frequencies: red dots: $u_{c1} = 2.54$, black dots: $u_{c2} = 1/|\beta|$. $\gamma = 10^{-3}$.

Appendix A

Some rational fractions in \sqrt{S}

A.1 Introduction

The use of a rational fraction in \sqrt{S}

$$Z(\sqrt{S}) = \frac{\sum_{m=0}^N b_m (\sqrt{S})^m}{\sum_{p=0}^P a_p (\sqrt{S})^p} \quad (\text{A.1})$$

has been proposed by Pintelon et al. [21, 22]. Some rational fraction in \sqrt{S} are studied below.

A.2 $\frac{1}{1 + \sqrt{S}}$

$$H(u) = \frac{1}{1 + \sqrt{i u}} \quad (\text{A.2})$$

$$\text{Re } H(u) = \frac{\sqrt{2}\sqrt{u} + 2}{2(u + \sqrt{2}\sqrt{u} + 1)}, \quad \text{Im } H(u) = -\frac{\sqrt{u}}{\sqrt{2}(u + \sqrt{2}\sqrt{u} + 1)} \quad (\text{A.3})$$

$$\begin{aligned} |H(u) - (1/2 + i/2)| &= \sqrt{(\text{Re } H(u) - 1/2)^2 + (\text{Im } H(u) - 1/2)^2} = \frac{\sqrt{2}}{2} \\ &\Rightarrow \text{circle, radius} = \frac{\sqrt{2}}{2} \end{aligned} \quad (\text{A.4})$$

Nyquist diagram: Fig. A.1.

A.3 $\frac{1}{1 + (\sqrt{S})^3}$

$$H(u) = \frac{1}{1 + (i u)^{(3/2)}} \quad (\text{A.5})$$

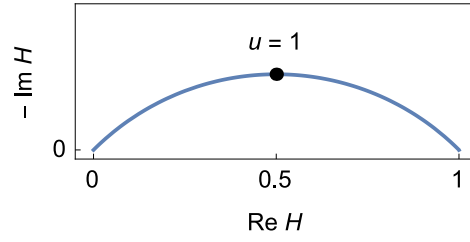


Figure A.1: Nyquist diagram of $\frac{1}{1 + \sqrt{i}u}$. One quarter circle. $u_c = 1$,
 $\text{Im } H(u_c) = -\frac{1}{\sqrt{2}(2 + \sqrt{2})}$.

$$\text{Re } H(u) = \frac{\sqrt{2}u^{3/2} - 2}{2(\sqrt{2}u^{3/2} - u^3 - 1)}, \quad \text{Im } H(u) = -\frac{u^{3/2}}{-2u^{3/2} + \sqrt{2}u^3 + \sqrt{2}} \quad (\text{A.6})$$

$$|H(u) - (1/2 - i/2)| = \sqrt{(\text{Re } H(u) - 1/2)^2 + (\text{Im } H(u) - 1/2)^2} = \frac{\sqrt{2}}{2}$$

$$\Rightarrow \text{circle, radius} = \frac{\sqrt{2}}{2} \quad (\text{A.7})$$

Remarkable frequencies

$$u_1 = \sqrt[3]{3 - 2\sqrt{2}}, \quad u_2 = 1, \quad u_3 = \sqrt[3]{2}, \quad u_4 = \sqrt[3]{3 + 2\sqrt{2}} \quad (\text{A.8})$$

Nyquist diagram: Fig. A.2.

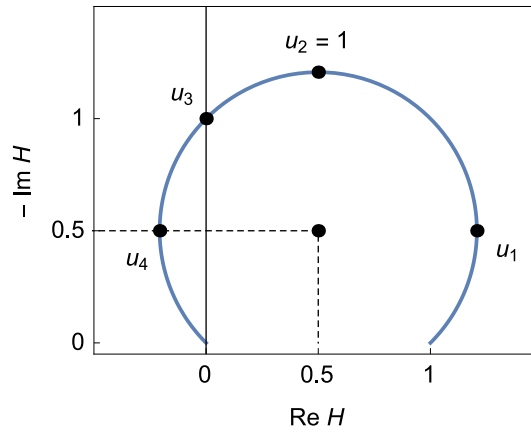


Figure A.2: Nyquist diagram of $\frac{1}{1 + (iu)^{3/2}}$. Three quarter circle.

A.4. $\frac{1 + (\sqrt{S})^2}{\sqrt{S}(1 + \alpha(\sqrt{S})^2)}$

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A.4 $\frac{1 + (\sqrt{S})^2}{\sqrt{S}(1 + \alpha(\sqrt{S})^2)}$

$$H(u) = \frac{1 + (\sqrt{iu})^2}{\sqrt{iu}(1 + \alpha(\sqrt{iu})^2)} \quad (\text{A.9})$$

$$\text{Re } H(u) = \frac{u(\alpha(u-1) + 1) + 1}{\sqrt{2}\sqrt{u}(\alpha^2 u^2 + 1)}, \quad \text{Im } H(u) = \frac{-u(\alpha u + \alpha - 1) - 1}{\sqrt{2}\sqrt{u}(\alpha^2 u^2 + 1)} \quad (\text{A.10})$$

Three different limiting cases

- $\alpha \ll 1$, Nyquist and Bode diagrams: Fig. A.3

$$u_1 = u_{\text{Im } H=0} = \frac{-\sqrt{\alpha^2 - 6\alpha + 1} - \alpha + 1}{2\alpha} \approx 1 \quad (\text{A.11})$$

$$u_2 = u_{\text{Im } H=0} = \frac{+\sqrt{\alpha^2 - 6\alpha + 1} - \alpha + 1}{2\alpha} \approx \frac{1}{\alpha} \quad (\text{A.12})$$

- $\alpha = 1$, $H(u) = \frac{1}{\sqrt{iu}}$

- $\alpha \gg 1$, Nyquist diagram: Fig. A.4

$$u_1 = u_{\text{Re } H=0} = \frac{-\sqrt{\alpha^2 - 6\alpha + 1} + \alpha - 1}{2\alpha} \approx \frac{1}{\alpha} \quad (\text{A.13})$$

$$u_2 = u_{\text{Re } H=0} = \frac{\sqrt{\alpha^2 - 6\alpha + 1} + \alpha - 1}{2\alpha} \approx 1 \quad (\text{A.14})$$

A.5 $\frac{1 - (\sqrt{S})^2}{\sqrt{S}(1 - \alpha(\sqrt{S})^2)}$

$$H(u) = \frac{1 - (\sqrt{iu})^2}{\sqrt{iu}(1 - \alpha(\sqrt{iu})^2)} \quad (\text{A.15})$$

$$\text{Re } H(u) = \frac{u(\alpha + \alpha u - 1) + 1}{\sqrt{2}\sqrt{u}(\alpha^2 u^2 + 1)}, \quad \text{Im } H(u) = \frac{u(\alpha + \alpha(-u) - 1) - 1}{\sqrt{2}\sqrt{u}(\alpha^2 u^2 + 1)} \quad (\text{A.16})$$

Three different limiting cases

- $\alpha \ll 1$, Nyquist diagram: Fig. A.5

$$u_1 = u_{\text{Re } H=0} = \frac{-\sqrt{\alpha^2 - 6\alpha + 1} - \alpha + 1}{2\alpha} \approx 1 \quad (\text{A.17})$$

$$u_2 = u_{\text{Re } H=0} = \frac{\sqrt{\alpha^2 - 6\alpha + 1} - \alpha + 1}{2\alpha} \approx \frac{1}{\alpha} \quad (\text{A.18})$$

- $\alpha = 1$, $H(u) = \frac{1}{\sqrt{iu}}$

- $\alpha \gg 1$.

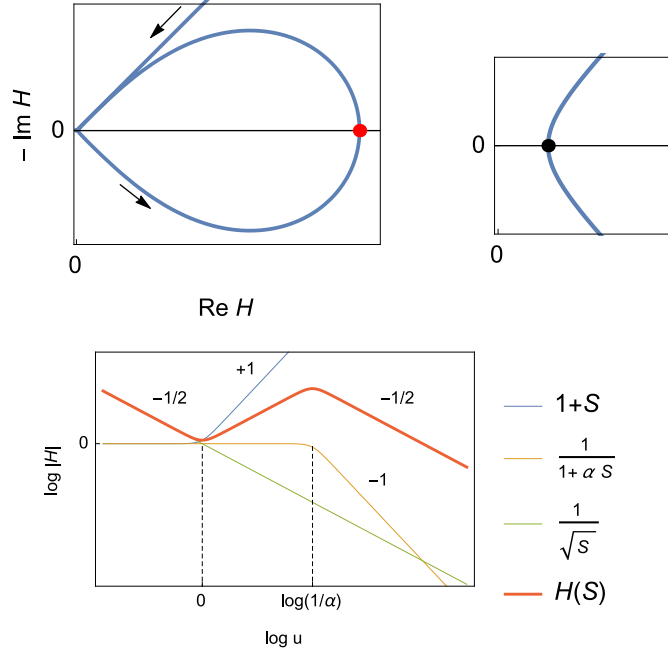


Figure A.3: Nyquist and Bode (modulus) diagram of $\frac{1 + (\sqrt{iu})^2}{\sqrt{iu}(1 + \alpha(\sqrt{iu})^2)}$. $a \ll 1$. Red dot: $u_2 \approx 1/\alpha$, black dot: $u_1 = 0 \approx 1$.

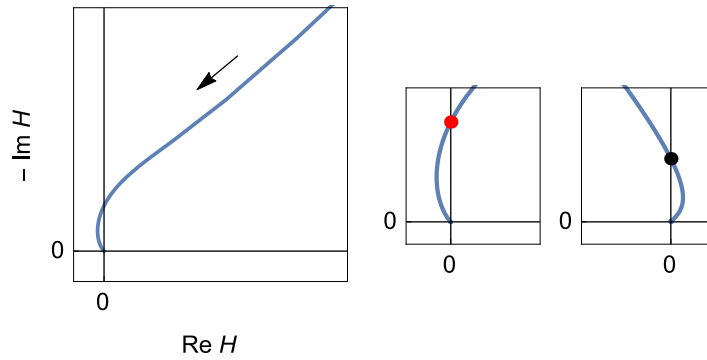


Figure A.4: Nyquist and Bode (modulus) diagram of $\frac{1 + (\sqrt{iu})^2}{\sqrt{iu}(1 + \alpha(\sqrt{iu})^2)}$. $a \gg 1$. Red dot: $u_{\text{Im } H=0} \approx 1/\alpha$, black dot: $u_{\text{Im } H=0} \approx 1$.

A.5. $\frac{1 - (\sqrt{S})^2}{\sqrt{S}(1 - \alpha(\sqrt{S})^2)}$

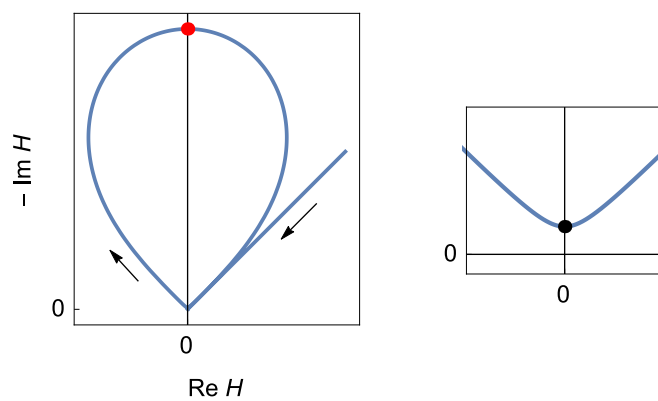


Figure A.5: Nyquist and Bode (modulus) diagram of $\frac{1 - (\sqrt{iu})^2}{\sqrt{iu}(1 - \alpha(\sqrt{iu})^2)}$. $a \ll 1$. Red dot: $u_2 \approx 1/\alpha$, black dot: $u_1 = 0 \approx 1$.

Appendix B

Impedance structure of reactions involving both adsorbed and soluble species

Table B.1: Impedance structure of reactions involving both adsorbed and soluble species. First order denominator.

Expression	Reaction	Impedance
$1 + \alpha \frac{\text{th } \sqrt{S}}{\sqrt{S}}$	(V-H) $A^+ + s + e^- \rightarrow A,s$ $A^+ + A,s + e^- \rightarrow A_2 + s$	Z_θ
$\frac{1 + \alpha \frac{\text{th } \sqrt{S}}{\sqrt{S}}}{1 + \beta S}$		

Table B.2: Impedance structure of reactions involving both adsorbed and soluble species. Second order denominator.

Expression	Reaction	Impédance
$1 + \alpha s + \beta \frac{\text{th } \sqrt{S}}{\sqrt{S}} + \gamma s \frac{\text{th } \sqrt{S}}{\sqrt{S}}$	(V-H) with desorption $A^+ + s + e^- \rightarrow A,s$ $A^+ + A,s + e^- \rightarrow A_2,s$ $A_2,s \rightarrow A_2 + s$	Z_θ
$\frac{1 + \alpha s + \beta \frac{\text{th } \sqrt{S}}{\sqrt{S}} + \gamma s \frac{\text{th } \sqrt{S}}{\sqrt{S}}}{1 + \delta S + \epsilon S^2}$		

Table B.3: Impedance structure of reactions involving both adsorbed and soluble species. Dénominateur : $1 + \alpha S + \beta \frac{\text{th} \sqrt{S}}{\sqrt{S}}$

Expressions	Reactions	Impédances
$\frac{1}{1 + \alpha S + \beta \frac{\text{th} \sqrt{S}}{\sqrt{S}}}$	(catalytic) $A^{2+} + B^- + s + e^- \rightarrow AB,s$ $AB,s + e^- \rightarrow A + B^- + s$ Hyp. $A^{2+}(0, t) = \text{cte}$	Z_θ
$\frac{\frac{\text{th} \sqrt{S}}{\sqrt{S}}}{1 + \beta s + \gamma S \frac{\text{th} \sqrt{S}}{\sqrt{S}}}$	(catalytic) $A^{2+} + B^- + s + e^- \rightarrow AB,s$ $AB,s + e^- \rightarrow A + B^- + s$ Hyp. $A^{2+}(0, t) = \text{cte}$	Z_{B^-}

Table B.4: Impedance structure of reactions involving both adsorbed and soluble species. Dénominateur : $1 + \beta S + \gamma S \frac{\text{th} \sqrt{S}}{\sqrt{S}}$.

Expressions	Reactions	Impedances
$\frac{1 + \alpha S}{1 + \beta S + \gamma S \frac{\text{th} \sqrt{S}}{\sqrt{S}}}$	(V-H)#2 $A^+ + s + e^- \rightarrow A,s$ $A,s + e^- \rightarrow A^- + s$	Z_θ
$\frac{1 + \alpha \frac{\text{th} \sqrt{S}}{\sqrt{S}}}{1 + \beta S + \gamma S \frac{\text{th} \sqrt{S}}{\sqrt{S}}}$	(E-EAR) $R + s \rightarrow O + s + n_1 e^-$ $A^- + s \leftrightarrow A,s + n_2 e^-$	Z_θ
$\frac{(1 + \alpha S) \frac{\text{th} \sqrt{S}}{\sqrt{S}}}{1 + \beta S + \gamma S \frac{\text{th} \sqrt{S}}{\sqrt{S}}}$	(E-EAR) $R + s \rightarrow O + s + n_1 e^-$ $A^- + s \leftrightarrow A,s + n_2 e^-$	Z_R
	(V-H)#2 $A^+ + s + e^- \rightarrow A,s$ $A,s + e^- \rightarrow A^- + s$	Z_{A^+}

Table B.5: Impedance structure of reactions involving both adsorbed and soluble species. Dénominateur : $1 + \beta s + \gamma \frac{\text{th} \sqrt{S}}{\sqrt{S}} + \delta S \frac{\text{th} \sqrt{S}}{\sqrt{S}}$.

Expression	Reaction	Impedance
$\frac{1 + \alpha \frac{\text{th} \sqrt{S}}{\sqrt{S}}}{1 + \beta S + \gamma \frac{\text{th} \sqrt{S}}{\sqrt{S}} + \delta S \frac{\text{th} \sqrt{S}}{\sqrt{S}}}$	(DP3) $M,s \rightarrow M^{2+} + s + 2 e^-$ $M,s + A^{2-} \rightarrow MA,s + 2 e^-$ $MA,s + B \rightarrow MAB + s$ Hyp. $A^{2-}(0, t) = A^{2-*}$	Z_s

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